

1. One method commonly employed to extract Doppler information in a form suitable for display on the PPI scope is with a [01D01]

- a. Power Amplifier
- b. A-Scope display
- c. Delay line canceller**
- d. Coherent oscillator

2. The characteristic feature of coherent MIT Radar is that the [01D02]

- a. Transmitted signal must be out of phase with reference signal in receiver
- b. The transmitted signal must be equal in the magnitude with reference signal
- c. The transmitted signal must be coherent with the reference signal in the receiver
- d. Transmitted signal must not be equal to reference signal in the receiver**

3. In the following which are produce, with time a butterfly effect on the 'A' scope [01M01]

- a. Fixed Targets
- b. PPI scope
- c. Moving Targets**
- d. Phase Detector

4. The stalo, coho and the mixer in which they are combined plus any low-level amplification are called the [01M02]

- a. Transmitter-Oscillator
- b. Transmitter-Exciter
- c. Receiver-Amplifier
- d. Receiver-exciter**

5. The Doppler frequency shift produced by a moving target may be used in a pulse radar to [01S01]

- a. Combine moving targets from desired stationery objects
- b. Determine the relative velocity of a target**
- c. Separate desired moving targets from desired stationery objects
- d. Determine the displacement of a target

6. To operate with unambiguous Doppler pulse repetition frequency is usually [01S02]

- a. Low
- b. Very low
- c. High**
- d. Very High

7. MTI stands for [01S03]

- a. Moving Transmitter Indicator
- b. Moving target interval
- c. Moving target indication**
- d. Modulation Transmitting Interval

8. Echoes from fixed targets [01S04]

- a. Vary in amplitude
- b. Vary in frequency
- c. Vary in pulse interval
- d. Remains constant**

9. The limitation of pulse MTI radar which do not occur with cw radar [02D01]

- a. Blind speeds**
- b. Delay lines
- c. Requires more operating powers
- d. Requires complex circuitry

10. The presence of blind speeds within the Doppler frequency band reduces the [02D02]

- a. Output of the radar
- b. Detection capabilities of the radar**
- c. Unambiguous range
- d. Ambiguous range

11. The capability of delay line canceller depends on the [02M01]

- a. Quality of signal
- b. Pulse Interval
- c. Quality of the medium used as the delay line.**
- d. Delay time of the delay line

12. The output of the MTI receiver pahse detector be quantized into a sequence of digital words by

using [02M02]

- a. Digital quantizer
- b. Digital Phase detector
- c. Digital delay lines**
- d. Digital filter

13. A transmitter which consists of a stable low power oscillator followed by a power amplifier is called [02S01]

- a. POMA
- b. MOPA**
- c. MTI Radar
- d. CW radar

14. A simple MTI delay line canceller is an example of a [02S02]

- a. Frequency domain filter
- b. High pass filter
- c. Active filter
- d. Time domain filter**

15. The delay line must introduce a time delay equal to the [02S03]

- a. Time interval
- b. Pulse repetition interval**
- c. Pulse width
- d. Phase shift

16. The delay line canceller [02S04]

- a. Rejects the ac component of clutter
- b. Rejects the dc component of clutter**
- c. Allows the ac as well as dc
- d. It rejects all components

17. In pulse MTI radar, Doppler is measured by [02S05]

- a. Continuous signals
- b. Discrete samples**
- c. Constant period
- d. Constant amplitude

18. The output of the two single delay line cancellers in cascade is the [02S06]

- a. Double of that from a single canceller
 b. U times that from a single canceller
c. Square of that from a single canceller
 d. Same as that from a single canceller
- 19. To operate MTI radar with high pulse repetition frequencies [03D01]**
 a. λ fp must be small
 b. λ fp must be unity
 c. λ fp must be zero
d. fp must be large
- 20. If the first blind speed were 600 knots, the maximum unambiguous range would be ----- at a frequency of 300MHz [03D02]**
 a. 140 nautical miles
 b. 600 knots
 c. 140 knots
d. 130 nautical miles
- 21. The maximum unambiguous range [03D03]**
 a. $R_{unamb} = c / 2$
 b. $R_{unamb} = (cT)2$
 c. $R_{unamb} = cT/4$
 d. $R_{unamb} = (cT/2)1/2$
- 22. To operate MTI radar at low frequencies [03M01]**
 a. λ fp must be small
 b. λ fp must be zero
c. fp must be large
 d. λ fp must be unity
- 23. The effect of blind speed can be significantly reduced in [03M02]**
 a. Pulse MTI radar
 b. Delay line canceller
c. Staggered - prf MTI
 d. Pulse canceller
- 24. The blind speeds are present in pulse radar because [03S01]**
 a. Doppler is measured by discrete samples at the prf
 b. Doppler is measured by continuous signal
 c. Doppler is assumed to be zero
 d. Doppler frequency remains constant
- 25. If the first blind speed is to be greater than the maximum radial velocity expected from the target, the product fp must be [03S02]**
 a. Small
 b. Zero
c. Large
 d. Infinity
- 26. The clutter-rejection notches may be widened by passing the output of the delay line canceller through a [03S03]**
 a. Coho
 b. Stalo
c. Second delay line canceller
 d. Pulse canceller
- 27. The frequency response of double delay line canceller is [03S04]**
 a. $4 \sin \pi f d 1 T$
 b. $4 \pi \sin \pi f d 1 T$
c. $4 \sin^2 \pi f d 1 T$
 d. $(2 \pi \sin \pi f d 1 T)^2$
- 28. MTI radar primarily designed for the detection of aircraft must usually operate with [03S05]**
 a. Unambiguous Doppler
 b. Unambiguous blind speed
c. Ambiguous Doppler
 d. Ambiguous range
- 29. The blind speeds of two independent radars operating at the same frequency will be different if their [04D01]**
 a. Amplitudes are different
 b. Blind speeds are different
c. Pulse repetition frequencies are different
 d. Pulse intervals are different
- 30. A disadvantage of the staggered prf is its inability to [04D02]**
 a. Cancel second-time around echoes
b. Cancel second-time around clutter echoes
 c. Provide variable prf
 d. Provide pulse to pulse incoherence
- 31. Second-time around clutter echoes can be removed by use of a [04M01]**
 a. Stalo
 b. Coho
c. Constant prf
 d. Delay canceller
- 32. The loss of range information and the collapsing loss may be eliminated by [04M02]**
 a. Sampling the range
 b. Shaping the range
c. Quantizing the range
 d. Keeping constant range
- 33. Range gating is a process of [04M03]**
 a. Sampling the range into various samples
b. Quantizing the range into small intervals
 c. Getting constant range
 d. Removing the range intervals
- 34. When the switching is pulse to pulse. It is known as a [04S01]**
 a. Delay canceller
b. Staggered prf
 c. MTI radar
 d. Pulse radar
- 35. Pulse to pulse coherence is provided by use of [04S02]**
 a. Stalo
 b. Coho
c. Constant prf
 d. Delay canceller
- 36. The output of the range gates is stretched in a circuit called [04S03]**
 a. Clutter rejection filter
 b. Clutter filter
c. Box car generator

d. Sampler

37. The clutter rejection filter is a [04S04]

a. Band stop filter

b. Bandpass filter

c. Lowpass filter

d. Highpass filter

38. The bandwidth of clutter rejection filter depends upon the extent of the [04S05]

a. Spectrum

b. Expected clutter spectrum

c. Filter characteristic

d. Clutter characteristic

39. By quantizing the range [05D01]

a. Loss of range Information is eliminated

b. Loss of range information is increased

c. Range becomes constant

d. Range becomes sampled

40. Clutter visibility factor provides [05D02]

a. Attenuation

b. Probabilities of detection and false alarm

c. Cancellation ratio

d. Decrease in clutter residue

41. In box car generator [05D03]

a. Output of range gates is quantized

b. Output of range gates is sampled

c. Output of range gates is stretched

d. Range gate output remains constant

42. The clutter power remaining at the output of MIT system is [05M01]

a. Clutter pulse

b. Clutter residue

c. Clutter Attenuation

d. Clutter power ratio

43. Amplitude limiting cannot be employed in [05M02]

a. Coherent MTI receiver

b. Pulse Doppler radar

c. Non-coherent MTI receiver

d. CW radar

44. It provides stated probabilities of detection and false alarm [05S01]

a. Clutter attenuations

b. Clutter ratio

c. Cancellation ratio

d. Clutter visibility factor

45. MTI radar which uses amplitude fluctuations is [05S02]

a. Coherent

b. Pulsed Doppler

c. Non coherent

d. CW radar

46. Constant prf is helpful to provide [05S03]

a. Pulse to pulse coherence

b. Stalo

c. Delay

d. Coho

47. Clutter residue is [05S04]

a. Clutter input power

b. Clutter power remaining at output of MTI system

c. Clutter output power ratio

d. Clutter output attenuation

48. By using constant prf [05S05]

a. Clutter factor is minimized

b. Clutter input power is increased

c. Clutter echoes can be removed

d. Range gate output can be sampled

49. This radar designates targets to the tracking radar by providing the coordinates where the targets are to be found. [06D01]

a. TWS radar

b. Sequential radar

c. Acquisition radar

d. CW radar

50. The antenna pattern commonly employed with tracking radar is the [06D02]

a. Symmetrical beam

b. Symmetrical Pencil Beam

c. Asymmetrical Pencil beam

d. Asymmetrical beam

51. One of method of obtaining the direction and magnitude of the angular error in one coordinate is

by alternately switching the antenna beam between two positions. This is called [06D03]

a. Lobe switching

b. Asymmetrical switching

c. Symmetrical switching

d. Sequential tracking

52. When the output from more than one radar are automatically combined to provide target tracks,

the processing is called [06S01]

a. ADIT

b. IDAT

c. GCA

d. PPI

53. A surveillance radar that provides target tracks is sometimes called a [06S02]

a. Track acquisition radar

b. Track-while-scan radar

c. Integrated ADT

d. ADT

54. Landing radars used for [06S03]

a. ADT

b. IADT

c. GCA

d. ADIT

55. ADIT stands for [06S04]

a. Automatic Decode Interlink Track

b. Automatic Decode Integrated Track

c. Automatic Detection and Integrated Track

d. Automatic Demodulation and Interlink Track

56. The difference between the target position and the reference direction is the [06S05]

a. Lobe error

b. Tracking error

c. Angular error

d. Sequential error

57. When the target is located along the reference direction [06S06]

- a. Lobe error is zero
- b. Lobe error is maximum
- c. Angular error is maximum

d. Angular error is zero

58. GCA stands for [06S07]

- a. General Control of Approach
- b. Ground Connection of Approach
- c. Ground Control of Approach**
- d. General Connection of Approach

59. In this technique the RF signals received from two offset antenna beams are combined so that

both the sum and the difference signals are obtained simultaneously [07D01]

- a. Monopulse**
- b. Pulse to pulse comparison
- c. Fixed lobing
- d. Sequential lobing

60. The mono pulse antenna must generate a sum pattern with [07D02]

- a. Minimum boresight gain
- b. Maximum boresight gain**
- c. Minimum pulse gain
- d. Maximum pulse gain

61. This is used to rotate continuously an offset antenna beam [07M01]

- a. rotating Feed
- b. rotational scanning
- c. conical scanning**
- d. nutating feed

62. The process of stretching the pulses before low-pass filtering is called [07M02]

- a. Sampling
- b. Quantizing
- c. Sample and hold**
- d. Detecting

63. A conical scan-on-receive-only tracking radar radiates a [07M03]

- a. Scanning transmit beam
- b. Scanning non-transmit beam
- c. Non scanning receive beam
- d. Non scanning transmit beam**

64. The angle between the axis of rotation and the axis of the antenna beam is called the [07S01]

- a. Lobe Angle
- b. Conical Angle
- c. Squint Angle**
- d. Rotation Angle

65. If the feed maintains the plane of polarization fixed as it rotates it is called a [07S02]

- a. Rotating Feed
- b. Fixed feed
- c. Nutating Feed**
- d. Flexible feed

66. When the antenna is an target, the line of sight to the target and the rotation axis coincide, and

the conical scan modulation is [07S03]

- a. Maximum
- b. Zero**
- c. Constant
- d. Infinity

67. Extracting the modulation imposed on a repetitive train of narrow pulses is called [07S04]

- a. Scanning
- b. Conical scanning
- c. Box caring**
- d. Sampling

68. LORO stands for [07S05]

- a. Lobe of radiation only
- b. Lobe on radiation only
- c. Lobe on receive only**
- d. Lobe on radar only

69. The difference in amplitudes in the several antenna positions was proportional to the [08D01]

- a. Angle of arrival
- b. Phase
- c. Angular error**
- d. Tracking accuracy

70. A tracking radar which operates with phase information is similar to an active interferometer and might be called a [08M01]

- a. Amplitude comparison monopulse radar
- b. Phase monopulse radar
- c. Simultaneous phase comparison radar**
- d. Hybrid Tracking

71. The antenna beams not offset in [08M02]

- a. Amplitude comparison monopulse radar
- b. Phase comparison monopole radar**
- c. Hybrid tracking radar
- d. Automatic tracking radar

72. The monopulse radar used [08S01]

- a. Single beam
- b. Time shared beam
- c. Two or more timeshared beams
- d. Two or more simultaneous beams**

73. High sidelobes are the result of [08S02]

- a. Sequential lobes
- b. Grating lobes**
- c. Simultaneous lobes
- d. Symmetrical lobes

74. Hybrid tracking system is a [08S03]

- a. Monopulse system
- b. Conical scan system
- c. Combination of monopulse and conical scan**
- d. Sequential lobing

75. In this technique, target amplitude fluctuations do not affect the tracking accuracy [08S04]

- a. Cono pulse system**
- b. Monopulse system

- c. Conical scan system
- d. Phase comparison monopulse

76. The sequential lobing and conical scan techniques used [08S05]

- a. Simultaneous beam
- b. Two or more simultaneous beams
- c. Single time shared antenna beam**
- d. Two or more time shared beam

77. The phase and amplitude comparison principles can be combined in a single radar to produce [08S06]

- a. Two dimensional angle tracking with four antenna beams
- b. One dimensional angle tracking with four antenna beams
- c. Two dimensional angle tracking with only two antenna beams**
- d. One dimensional angle tracking with only two antenna beams

78. The problem of high side lobes can be reduced by [08S07]

- a. Reducing antenna apertures
- b. Reducing the angular error
- c. Overlapping the antenna apertures**
- d. Reducing angle sensitivity

79. This scan covers an angular search volume with circular symmetry [09D01]

- a. Palmer scan
- b. Spiral scan
- c. Helical scan
- d. Circular scan**

80. This scan suffer from the disadvantage that all parts of the scan volume do not receive the same energy unless the scanning speed is varied during the scan cycle [09D02]

- a. Helical scan
- b. Palmer scan
- c. Lowpass filter
- d. Conical scan**

81. The range gating [09M01]

- a. Isolates on target, excluding targets at other ranges**
- b. Does not permits the box cargenerator
- c. Antenna is continuously rotated
- d. Can cause noise in an analogous manner

82. The range gating [09M02]

- a. Isolates on target, excluding targets at other ranges**
- b. Does not permits the box cargenerator
- c. Antenna is continuously rotated
- d. Can cause noise in an analogous manner

83. These are used to obtain hemispheric coverage with a pencil beam [09M03]

- a. Spiral and palmer scan
- b. Spiral and helical scan
- c. Helical and nodding scan**
- d. Helical and conical scan

84. The random wandering of the apparent radar reflecting center gives rise to [09S01]

- a. Target glint**

- b. Random noise
- c. Radar fluctuations
- d. Center fluctuations

85. This type of noise depends on the length of the target and its shape [09S02]

- a. Random Noise
- b. Gating Noise
- c. Range-tracking noise**
- d. Radar-tracking noise

86. In this scan the antenna is continuously rotated in azimuth while it is simultaneously raised or

lowered in elevation [09S03]

- a. Plamer scan
- b. Spiral Scan
- c. Circular scan
- d. Helical scan**

87. It consists of a rapid circular scan about the axis of the antenna combined with a linear movement of the axis of the rotation [09S04]

- a. Palmer Scan**
- b. Spiral Scan
- c. Helical scan
- d. Circular scan

88. When the axis of rotation is held stationary, the palmer scan reduces to the [09S05]

- a. Stationary scan
- b. Spiral scan
- c. Conical scan**
- d. Helical scan

89. It is used with high-finding radars [09S06]

- a. Helical scan
- b. Nodding scan**
- c. Palmer scan
- d. Conical scan

90. When the target is being tracked, the signal-to-noise ratio available from the monopulse radar is [10D01]

- a. Less than that of a conical scan radar
- b. Equal to that of a conical scan radar
- c. Greater than that of a conical scan radar
- d. Zero**

91. This radar first makes its angle measurement and then integrates a number of pulses to obtain

the required signal-to-noise ratio and to smooth the error [10D02]

- a. Conical scan radar
- b. Monopulse radar
- c. Sequential lobing**
- d. Helical scan radar

92. This radar integrates a number of pulses first and then extracts the angle measurement [10M01]

- a. Conical scan radar**
- b. Monopulse radar
- c. Helical scan radar

d. Spiral scan radar

93. With the monopulse tracker it is possible to obtain a measure of the angular error in two coordinates on the basis of [10M02]

- a. Four pulses
- b. Single Pulses
- c. Dual Pulses**
- d. Many Pulses

94. The side lobe levels are higher than desired in this radar [10S01]

- a. Sequential lobbing
- b. Conical scan
- c. Amplitude comparison monopulse
- d. Phase comparison monopulse**

95. It suffers less loss and the antenna and feed systems are usually less complex [10S02]

- a. Sequential lobbing
- b. Conical scan**
- c. Helical scan
- d. Spiral scan

96. This radar having more tracking accuracy [10S03]

- a. Conical scan radar
- b. Helical scan radar
- c. Monopulse radar**
- d. Sequential lobbing

97. It is the preferred technique for precision tracking [10S04]

- a. Conical scan
- b. Monopulse radar**
- c. Circular scan
- d. Spiral scan

98. It is not degraded by amplitude fluctuations [10S05]

- a. Conical scan radar
- b. Monopulse radar**
- c. Helical scan
- d. Spiral scan

99. It is less costly and less complex [10S06]

- a. Conical scan radar**
- b. Surveillance radar
- c. Monopulse radar
- d. Phased array radar

100. North filter frequency response function is [11D01]

- a. $H(f) = GaS^*(f) \exp(j2\pi ft)$
- b.**
- c.
- d. $H(f) = GaS^*(f) \exp(-j2\pi ft)$

101. The matched filter may also be specified by [11D02]

- a.
- b.
- c.
- d.

102. Peak signal-to-mean noise ratio of the matched filter [11D03]

- a.
- b.**
- c.
- d.

103. The amplitude spectrum of the matched filter is [11M01]

- a. Obtained by its frequency response
- b. Negative of the amplitude spectrum of signal
- c. Same as the amplitude spectrum of signal**
- d. Depends on phase spectrum of signal

104. The noise power per hertz of bandwidth, N_0 is equal to [11M02]

- a.
- b.
- c.
- d.

105. A network whose frequency response function maximizes the output peak-single-to mean Noise ratio is called a [11S01]

- a. Envelope Detector
- b. Matched Detector
- c. Matched Filter**
- d. Optimum filter

106. If the band width of the receiver passband is wide compared with that occupied by the signal energy [11S02]

- a. Extraneous noise is introduced**
- b. Noise is reduced
- c. Increases the signal to noise ratio
- d. Frequency response is improved

107. If the receiver bandwidth is narrower than the bandwidth occupied by the signal [11S03]

- a. Extraneous noise is introduced
- b. Noise energy is reduced**
- c. Increases the signal-to-noise ratio
- d. Frequency response is improved

108. When there is optimum bandwidth [11S04]

- a. Signal to noise ratio is minimum
- b. Signal to noise ratio is maximum**
- c. Noise energy is maximum
- d. Noise energy is reduced

109. Phase spectrum of the matched filter is [11S05]

- a. Inversely proportional to frequency
- b. Negative of the phase spectrum of signal**
- c. Same as the phase spectrum of signal
- d. Proportional to amplitude

110. The cross correlation function $R(t)$ of two signals $y(t)$ and $s(t)$ each of finite duration is defined as [12D01]

- a.
- b.
- c.
- d.

111. The matched filter forms the [12D02]

- a. Cross correlation between transmitted signals
- b. Cross correlation between signal corrupted by noise and replica of transmitted signal
- c. Correlation between signal corrupted by noise and replica of transmitted signal**

d. Cross correlation between received signal corrupted by noise and a replica of the transmitted signal

112. The output $y_0(t)$ matched filter with impulse response $h(t)$ when the input is $y_{in}(t) =$

$s(t)+n(t)$ is [12D03]

- a.
- b.
- c.
- d.

113. When the input noise is stationary and white the peak signal to mean noise ratio is

[12M01]

- a.
- b.
- c.
- d.

114. The maximum ratio of the peak signal power to the mean noise power is [12M02]

- a.
- b.
- c.
- d.

115. The Output of the matched filter is [12S01]

- a. Replica of the input signal
- b. Proportional to input signal to noise ratio
- c. **Not a replica of the input signal**
- d. Not preserving the shape of the input signal

116. The output of the matched filter proportional to [12S02]

- a. Input signal
- b. Input signal correlated with a replica of transmitted signal
- c. Input signal correlated with a replica of transmitted signal except for time delay t_1
- d. **Input signal cross correlated with a replica of the transmitted signal except for the time delay**

117. The auto correlation function of a rectangular pulse of width τ is a [12S03]

- a. Square whose base is of width τ
- b. Triangular whose base is of width τ
- c. **Triangular whose base is of width 2τ**
- d. Square whose base is of width 2τ

118. If the input signal $y_{in}(t)$ were the same as the signal $s(t)$ for which the matched filter was designed, the output would be the [12S04]

- a. Correlation function
- b. Cross correlation function
- c. **Auto correlation function**
- d. Replica of the transmitted signal

119. When the input signal $s(t)$ is a rectangular sine wave pulse the output peak-signal to mean noise ratio is [12S05]

- a. Same as signal to noise power ratio
- b. Twice the signal to noise power ratio
- c. Same as average signal to noise power ratio

d. Twice the average signal to noise power ratio

120. This equation describes the output $y_0(t)$ of the matched filter as the [13D01]

- a. Correlation between the input signal and transmitted signal
- b. **Cross correlation between the input signal and a delayed replica of the transmitted signal**
- c. Frequency response function
- d. Correlation function

121. This requires a longer search time [13D02]

- a. Test for presence of a target at a single time delay
- b. **Test for presence of targets at time delays founds by varying T_r**
- c. Test for target at T_r
- d. Test for target at $(t-T_r)$

122. The cross correlation receiver tests for the presence of a target at [13M01]

- a. **Only a single time delay**
- b. Various time delays
- c. Two time delays T_1 and T_2
- d. Two time delays by means of the mixer

123. In cross correlation detection the input signal $y(t)$ is multiplied by a delayed replica of the transmitted signal $s(t-T_r)$ and the product is passed through [13M02]

- a. Mixer to get integrated output
- b. Delay T_r
- c. **Lowpass filter to perform the integration**
- d. High pass filter to perform the differentiation

124. The matched filter receiver can be replaced by this receiver that performs the same

Mathematical

operation [13S01]

- a. Correlation receiver
- b. **Cross correlation receiver**
- c. Super Hetrodyne receiver
- d. Non matched filter receiver

125. By adding parallel channels search time can be [13S02]

- a. Increased
- b. **Reduced**
- c. Remains constant
- d. Zero

126. In correlation detection mixer produces [13S03]

- a. Integrated output of input
- b. **Product of input and delayed replica of transmitted signal**
- c. Delayed replica of input signal
- d. Delayed replica of transmitted signal

127. In proportion to the number of time delay intervals T_r that are to be tested. The play back speed

is [13S04]

- a. Decreased
- b. **Increased**
- c. Independent of time delay
- d. Twice that of T_r

128. Cross correlation receiver and the matched filter receiver are [13S05]

- a. Not equivalent
- b. Equivalent practically
- c. Equivalent mathematically**
- d. Not equivalent mathematically

129. The frequency response function of the linear time invariant filter [13S06]

- a. Minimizes that output peak signal to noise ratio
- b. Maximizes the output peak signal to mean noise ratio**
- c. Doubles the output peak signal to noise ratio
- d. Minimizes the output peak signal to mean noise ratio

130. The maximum efficiency of the single tuned filter occurs for [14D01]

- a. B = 0.4**
- b. BT = 0.6
- c. BT = 1.37
- d. BT = 0.72

131. For Gaussian pulse input, the maximum efficiency of the Gaussian filter occurs for

[14D02]

- a. B = 0.44**
- b. BT = 0.67
- c. BT = 1.37
- d. BT = 0.72

132. In matched filter the measure of efficiency is taken as [14M01]

- a. Peak signal to noise ratio from the matched filter divided by the peak signal to noise ratio from matched filter
- b.
- c. Peak signal to noise ratio from the non-matched filter divided by the peak to noise ratio from matched filter**
- d.

133. The maximum efficiency of the rectangular filter occurs (when input is rectangular pulse) for

[14M02]

- a. B = 1.37**
- b. BT = 0.72
- c. BT = 0.44
- d. BT = 0.67

134. In Gaussian pulse input rectangular filter loss in SNR compared with matched filter

[14M03]

- a. 0.49dB**
- b. 0.85dB
- c. 0dB
- d. 0.56dB

135. For rectangular pulse input Gaussian filter have loss in SNR compared with matched filter

[14S01]

- a. 0.49dB**
- b. 0.88dB
- c. 0.56dB
- d. 0.5dB

136. The loss in SNR incurred by use of the non matched filter is [14S02]

- a. large
- b. Small**
- c. Zero
- d. Same as matched filter

137. If filter consists of S cascaded single tuned stages with rectangular input, loss in SNR

compared with matched filter is [14S03]

- a. 0.88 dB
- b. 0.49 dB
- c. 0.5 dB**
- d. 0.9 dB

138. The loss in SNR incurred by use of these filters is small [14S04]

- a. Matched filters
- b. Non matched filters**
- c. Single tuned filters
- d. Gaussian filters

139. If filters consisting of 2 cascaded single tuned stages with rectangular pulse input loss in

SNR

compared with matched filter is [14S05]

- a. 0.88 dB
- b. 0.49 dB
- c. 0 dB
- d. 0.56 dB**

140. For white noise non white noise matched frequency response function reduces to [15D01]

- a. $H(f) = Ga S^*(f) \exp(j2\pi ft1)$
- b. $H(f) = Ga S^*(f) \exp(-j2\pi ft1)$**
- c. $H(f) = Ga S(f) \exp(-j2\pi ft1)$
- d. $H(f) = Ga S(f) \exp(j2\pi ft1)$

141. The non white noise matched filter can be considered as the [15D02]

- a. Cascade of whitening and non whitening filters
- b. Cascade of whitening and matched filters**
- c. Cascade of non whitening filters
- d. Cascade of matched, non matched filters

142. When NWN matched filter can be considered as cascade of two filters. The first filter is

[15D03]

- a. Whitening filter**
- b. Non whitening filter
- c. Gaussian filter
- d. Matched filter

143. When noise is non white, this filter maximizes output signal to noise ratio [15D04]

- a. Whitening filter
- b. NWN Matched filter**
- c. Gaussian filter
- d. Matched filter

144. If the input power spectrum of the interfering noise is given by $[N_i(f)]^2$ the frequency

response

function of the filter which maximizes the output signal to noise ratio is [15M01]

- a.
- b.

- c.
d.
- 145. When the noise is non white, non white noise matched filter [15M02]**
- Minimizes the output SNR
 - Maximizes the output SNR**
 - Doubles the output SNR
 - Doubles the output SNR divided by F_0
- 146. For white noise the input power spectrum of the interfering noise is [15S01]**
- Constant**
 - Zero
 - Increases
 - Decreases
- 147. The whitening filter acts to make [15S02]**
- Noise spectrum is zero
 - Noise spectrum uniform**
 - Noise spectrum non uniform
 - Finite noise spectrum
- 148. When NWN matched filter can be considered as cascade of two filters. The second filter is [15S03]**
- Whitening filter
 - Non whitening filter
 - Gaussian filter
 - Matched filter**
- 149. If the spectrum of the noise accompanying the signal was assumed to be white it indicates that the matched filter characteristic was [15S04]**
- Dependent of frequency
 - Independent of frequency**
 - Improved
 - Remains constant
- 150. The noise figure F_n of a linear network is defined as [16D01]**
- a.
 - b.
 - c.
 - d.
- 151. The output noise from two circuits in cascade is N_0 , the noise figure F_0 is [16D02]**
- a.
 - b.
 - c.
 - d.
- 152. If the receive effective noise temperature is T_e , then operating noise temperature is [16D03]**
- $T_0 F_s$**
 - T_0 / F_s
 - $T_0 / (1 + F_s)$
 - $T_0 / (1 - F_s)$
- 153. Effective noise temperature is defined as [16M01]**
- $T_e = (F_n - 1)T_0$**
 - $T_e = F_n T_0$
 - $T_e = F_n - 1 / T_0$
 - $T_e = (F_n + 1)T_0$
- 154. The noise figure of 'N' networks in cascade may be shown to be [16S01]**
- $F_0 = F_1 + F_2 + F_3 + \dots + F_n - 1$
 - b.
 - c.
 - d.
- 155. It is defined as the effective noise temperature of the receiver system including the effects of antenna temperature [16S02]**
- Effective noise temperature
 - Noise figure
 - System noise temperature**
 - Antenna noise temperature
- 156. The effective noise temperature of receiver consisting of a number of networks in cascade [16S03]**
- a.
 - b.
 - c.
 - d.
- 157. System noise temperature is given by [16S04]**
- $T_s = T_a + T_f$
 - $T_s = T_a + T_e$**
 - $T_s = T_a - T_f$
 - $T_s = T_a - T_e$
- 158. The expression for noise figure in terms of additional noise introduced by network it self is [16S05]**
- a.
 - b.
 - c.
 - d.
- 159. The noise figure is commonly expressed in [16S06]**
- Hertz
 - Decibels**
 - Hz/W
 - W/Hz
- 160. When a long persistence is needed the required phosphor is [17D01]**
- P7
 - P19**
 - P1
 - P39
- 161. This is universally used as the radar display [17M01]**
- Raw Video
 - Blip display
 - CRT display**
 - Synthetic display
- 162. Where no persistence is needed as when the frame time is less than the response time of the eye. This phosphor is commonly used [17M02]**
- P7
 - P19
 - P1**
 - P39

163. It is appropriate for PPI presentations where the frame times are several seconds

[17M03]

- a. P19
- b. P39
- c. P29
- d. P7

164. When the display is connected directly to the video output of the receiver the information displayed is called [17S01]

- a. Raw video
- b. Synthetic video
- c. CRT video
- d. Blip video

165. In this type of video, receiver video output is processed by automatic detection and tracking

processor [17S02]

- a. Synthetic Video
- b. Raw Video
- c. CRT Video
- d. Blip Video

166. An intensity modulated rectangular display with azimuth angle indicated by horizontal coordinate

and elevation angle by the vertical coordinate is [17S03]

- a. A Scope
- b. B Scope
- c. C Scope
- d. R Scope

167. This is the device that allows a single antenna to serve both the transmitter and the receiver

[17S04]

- a. Dual Pulse generator
- b. Duplexer
- c. Multiplexer
- d. Demultiplexer

168. This is based on the short-slot Hybrid function [17S05]

- a. Balanced Demodulator
- b. Balanced Duplexer
- c. Unbalanced Duplexer
- d. ATR Tube

169. The power handling capability is greater in the [17S06]

- a. Balanced Duplexer
- b. Branch Duplexer
- c. ATR Tube
- d. Receiver protector

170. Phased array is made up of [18D01]

- a. Non radiating antennas
- b. Radiating Antennas
- c. Loop Arrays
- d. Non Resonant arrays

171. The linear array generates a fan beam such that radiation of pattern [18D02]

- a. Perpendicular to array

- b. Parallel to array
- c. Lies in same plane as array
- d. Not lies in array plane

172. It offers separation of transmitter and receiver without need for conventional duplexer configurations [18M01]

- a. ATR Tube
- b. Ferrite Circulator
- c. TR Tube
- d. Diode Limiter

173. The ferrite circulator with receiver protector is attractive for radar applications because of its

[18M02]

- a. Narrow Bandwidth
- b. Long life
- c. Solid state configuration
- d. Low VSWR

174. The ferrite circulator have [18S01]

- a. Less life time
- b. Narrow bandwidth
- c. Wider bandwidth
- d. Complex designing

175. The phased array is a [18S02]

- a. Directive Antenna
- b. Resonant Antenna
- c. Non Resonant Antenna
- d. Non Directive Antenna

176. It consists of elements arranged in a straight line in one dimension [18S03]

- a. Non linear array
- b. Loop Array
- c. Linear Array
- d. Planar Array

177. It is a two dimensional configuration [18S04]

- a. Non linear array
- b. Loop Array
- c. Linear Array
- d. Planner Array

178. The VSWR is a measure of the amount of power [18S05]

- a. Incident by Antenna
- b. Radiated by Antenna
- c. Reflected by Antenna
- d. Isolated by Antenna

179. It is a good absorber device as compared to gas tube TR [18S06]

- a. Reflective limiter
- b. Circulator
- c. Ferrite limiter
- d. TR limiter

180. When directive elements are used, the resultant array antenna radiation pattern is

[19D01]

- a. $G(\theta) = G_e(\theta) G_a(\theta)$
- b. $G(\theta) = G_e(\theta) + G_a(\theta)$
- c. $G(\theta) = G_e(\theta) - G_a(\theta)$
- d. $G(\theta) = 2G_e(\theta) + G_a(\theta)$

181. If the radiation patterns in the two principal planes are $G_1(\theta_e)$ and $G_2(\theta_a)$ the two dimensional

antenna pattern is [19D02]

- a. $G_1(\theta_e) + G_2(\theta_a)$
- b. $2G_1(\theta_e) + G_2(\theta_a)$
- c. $G_1(\theta_e) G_2(\theta_a)$**
- d. $G_1(\theta_e)/G_2(\theta_a)$

182. Grating lobes caused by a widely spaced array be eliminated with [19M01]

- a. Directive elements**
- b. End fire arrays
- c. Array elements
- d. Lobe factors

183. Radiate little or no energy in the directions of the undesired lobes [19M02]

- a. Non radiative elements
- b. Non directive elements
- c. Directive elements**
- d. Non resonant elements

184. An array whose elements are distributed on a non planar surface is called [19S01]

- a. Electronically scanned array
- b. Conformal Array**
- c. Linear array
- d. Loop array

185. The array factor has also been called [19S02]

- a. Lobe factor
- b. Radiation factor
- c. Space factor**
- d. Element factor

186. This is the pattern of an array composed of Isotropic elements [19S03]

- a. Directive factor
- b. Lobe factor
- c. Element factor
- d. Space factor**

187. As the beam is scanned off the broad side direction. The half power beam width in the

plane of scan [19S04]

- a. Decreases
- b. Remains constant
- c. Reduced to zero
- d. Increases**

188. If the same phase is applied to all elements the relative phase difference between adjacent elements is [19S05]

- a. 90°
- b. 180°
- c. 45°
- d. zero**

189. If the spacing between antenna elements is $\lambda/2$, the pattern of uniformly illuminated array is [19S06]

- a. Similar to pattern of continuously illuminated uniform array**
- b. Different from pattern of continuously illuminated uniform array

- c. Same as pattern of non uniform array
- d. Similar to pattern of discontinuous illuminated array

190. In order to position the main beam of the radiation pattern at θ_0 the relative phase shift between

adjacent elements of the array must be [20D01]

- a.
- b.**
- c.
- d.

191. In this array the energy to be radiated is divided between the elements by a power splitter [20D02]

- a. Serial feed array
- b. Parallel fed array**
- c. Corporate feed
- d. Circular feed

192. The maximum phase change required of each phase shifter in the parallel feed array is [20M01]

- a. π Radians
- b. Radians
- c. Many times 2π radians**
- d. 2π radians

193. A two dimensional parallel feed array of MN elements requires [20M02]

- a. M+N separate control signals
- b. control signals
- c. M+N-2 separate control signals**
- d. separate control signals

194. When a series of power splitters are used to create a tree like structure is called [20S01]

- a. Series feed
- b. Parallel feed
- c. Corporate feed**
- d. Tree feed

195. In the series feed where the signal is fed from one end the position of the beam will [20S02]

- a. Remains constant
- b. Vary with signal strength
- c. Vary with frequency**
- d. Vary with phase shift

196. Each phase shifter in the series fed linear array has [20S03]

- a. Phase shift - 37π Radians
- b. Same value of phase shift**
- c. Phase shift greater than 2π radians
- d. Variation in phase with frequency

197. Phase shifter in a series feed array must be of [20S04]

- a. Higher loss compared to parallel feed array
- b. Lower loss compared to parallel feed array**
- c. Loss same as that of parallel feed array
- d. Zero

198. The proper phase change for beam steering is introduced by [20S05]

- a. Power splitters
- b. Phase shifters**

- c. Phase splitters
- d. Corporate feed

199. In the parallel feed array energy to be radiated is [20S06]

- a. Divided by a power splitter
- b. Obtained by phase shifter**
- c. Obtained by corporate feed

- d. Divided by a phase splitter